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# EBG-ANTENNA ANALYSIS : UNIFICATION OF FREQUENCY AND ANGULAR DEPENDENCIES

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**ABSTRACT :** In this paper, we present an original method to study planar EBG based antennas. From this method, which unifies the frequency and angular dependencies of the EBG structure, we obtain an analytical expression of the directivity of the structure versus frequency. To our knowledge, these results concerning EBG structures have never been presented.

## I. INTRODUCTION

Several recent works [1][2][3][4] deal with directive antennas based on EBG (Electronic Band Gap) structures. In these works the directivity characteristic is generally studied numerically. The purpose of this presentation is to establish an analytical expression and to show that the frequency and spatial behaviors of EBG structures are tightly interrelated. To this end we propose a new simple method for analyzing EBG-Antenna. Inspired from the antenna array theory [5], this method can allow, for example, to obtain the radiation pattern of an EBG-Antenna from the frequency curves. We will consider first a point source inside multiple Partially Reflecting Surfaces (PRS). Then we will study in more details the case of one PRS on each side of the point source, estimating the directivity and the performance of this EBG-Antenna. Finally, we will focus on a structure composed of a ground plane in the same plane of the source.

## II. GENERAL EQUATIONS

Let us consider an ideal point source inside a periodic structure composed of  $2n$  identical partially reflecting surfaces (Figure 1). In practice, the source can be a dipole as in ref. [2][3]. The source creates plane waves uniformly propagating in all  $\theta$  directions. Let  $(r)$  and  $(t)$  be respectively the complex reflection and transmission coefficients of a single layer ( $r=|r|\exp(j\varphi_r)$ ,  $t=|t|\exp(j\varphi_t)$ ).

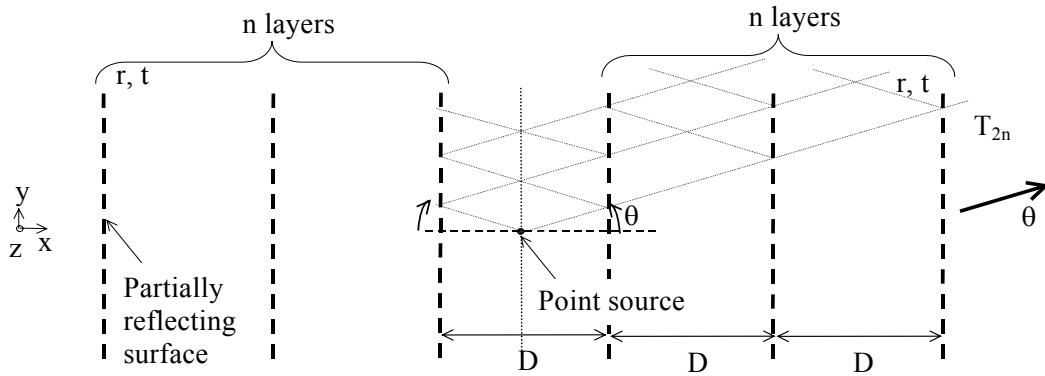


Figure 1 – Point source inside multiple ( $2n$ ) Partially Reflecting Surfaces (PRS).

We have calculated the transmission coefficient  $T_{2n}(\psi)$  of the structure due to two symmetrical plane waves at a given angle  $\theta$  (Figure 1). We have shown [6] that the transmission coefficient  $T_{2n}(\psi)$  can be written in a simple closed form :

$$T_{2n}(\psi) = \frac{t_n(\psi) \exp(-j\psi/2)}{1 - r_n(\psi) \exp(-j\psi)} \quad (1)$$

$t_n$  and  $r_n$  are evaluated by recursive equations from  $(r, t)$  [6] :

$$t_n(\psi) = \frac{t_{n-1}(\psi) \cdot t \cdot \exp(-j\psi)}{1 - r_{n-1}(\psi) \cdot r \cdot \exp(-j2\psi)} \quad (2)$$

$$r_n(\psi) = r_{n-1}(\psi) + \frac{t_{n-1}(\psi)^2 \cdot r \cdot \exp(-j2\psi)}{1 - r_{n-1}(\psi) \cdot r \cdot \exp(-j2\psi)} \quad (3)$$

$$t_1(\psi) = t, \quad r_1(\psi) = r, \quad (4)$$

$$\text{and} \quad \psi = kD \cos(\theta) \quad (5)$$

$k$  is the free space wave number and  $D$  is the distance between two consecutive partially reflecting surfaces (Figure 1).

It is important to note that the expression (1) unifies the frequency and incidence angle dependencies in a unique variable  $\psi = kD \cos(\theta)$ . Another point is that  $T_{2n}(\psi)$  is a periodic function of  $\psi$  with period  $2\pi$ . The knowledge of one period ( $0 \leq \psi \leq 2\pi$ ) is sufficient to extract the frequency curves ( $0 \leq f < \infty$ ) and/or the radiation pattern ( $-\pi/2 \leq \theta \leq \pi/2$ ).

In Figure 2, we give an example of  $|T_{2n}(\psi)|$  for  $|r|=0.7 \cdot \exp(j2.35)$ ,  $|t|=0.71 \cdot \exp(j0.78)$  [6] (for  $n=1, 2, 3$ ). To simplify the study, we consider that  $(r, t)$  don't depend on frequency or on incidence angle.  $|T_{2n}(\psi)|$  contains two types of pass-bands : pass-bands with high picks, and pass-bands with picks limited to one. The values of  $|T_{2n}(\psi)|$  higher than one are explained physically in ref. [6][7]. It does not correspond directly to antenna directivity as it was used in [8][9].

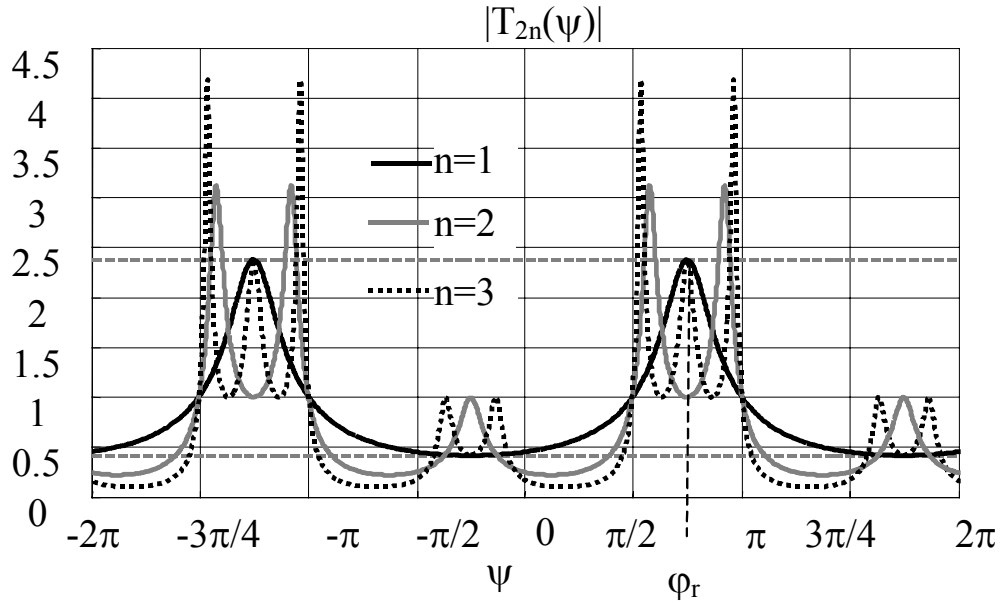


Figure 2 – Transmission coefficient versus  $\psi$  for  $n=1, 2, 3$  ( $|r|=0.7 \cdot \exp(j2.35)$ ,  $|t|=0.71 \cdot \exp(j0.78)$ ).

Now it's easy to pass from the  $\psi$  domain to the frequency domain (for a given angle  $\theta$ ) or to the angular domain (for a given frequency), by using the classical method of antenna array theory [5].

### III. STRUCTURE WITH ONE LAYER ON EACH SIDE OF THE SOURCE

#### III. 1 Extraction of frequency curve and radiation pattern

We consider one PRS on each side of the source (Figure 3). For this simple structure, we will show how to pass from the  $\psi$  domain to the frequency or angular domains. We will also evaluate the directivity of this structure and calculate the frequency which gives the maximum of directivity.

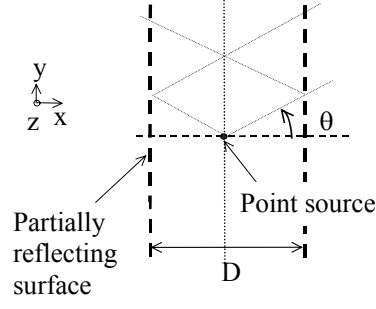


Figure 3 – Point source inside two partially reflecting surfaces

The magnitude of the transmission coefficient  $T_2(\psi) = \frac{t \cdot \exp(-j\psi/2)}{1 - r \cdot \exp(-j\psi)} = \frac{t \cdot \exp(-j\psi/2)}{1 - |r| \cdot \exp(-j(\psi - \phi_r))}$  is given in

Figure 4. To pass from the  $\psi$  domain to the frequency domain we use the following transformation :  $f = \frac{\psi \cdot c}{2\pi D \cos(\theta)}$ ,

which we wrote for  $\theta=0^\circ$   $f = \frac{\psi \cdot c}{2\pi D}$ .  $c$  is the speed of light. The resonance frequency  $f = f_0 = \frac{\phi_r \cdot c}{2\pi D}$  correspond to  $\psi = \phi_r$ .  $|T_2|$  is plotted versus the frequency  $f$  in the second abscissa of Figure 4.

To pass from the  $\psi$  domain to the angular domain we use  $\theta = \pm \arccos\left(\frac{\psi}{kD}\right)$ . For  $f=f_0$ ,  $\theta = \pm \arccos\left(\frac{\psi}{\phi_r}\right)$ . We use a

graphical method [5] as it is shown in Figure 4. Note that the well known “visible region” in the array theory starts from the 0 frequency and finishes at the frequency considered.

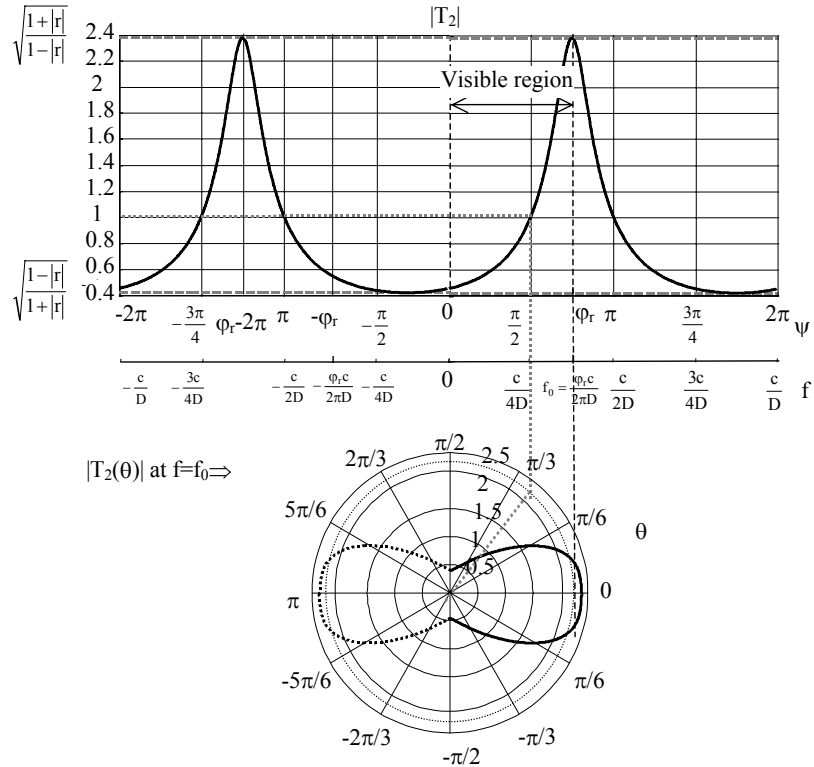


Figure 4 –  $|T_2|$ , transformations from the  $\psi$  domain to the frequency domain (at  $\theta=0^\circ$ ) and to the angular domain (at  $f=f_0$ )

It's interesting to note that the lower and upper envelopes of the transmission coefficient  $|T_2|$  are given by [6] :

$$|T_2|_{\max} = \frac{|t|}{1-|r|} = \frac{\sqrt{1-|r|^2}}{1-|r|} = \sqrt{\frac{1+|r|}{1-|r|}}, \quad |T_2|_{\min} = \frac{|t|}{1+|r|} = \frac{\sqrt{1-|r|^2}}{1+|r|} = \sqrt{\frac{1-|r|}{1+|r|}} = \frac{1}{|T_2|_{\max}} \quad (6)$$

Note that  $|T_2|_{\max}$  should not be considered as the gain enhancement of the structure as it is proposed in ref. [8][9]. We will see later how to calculate the directivity and then the true performance of the structure.

Figure 5 gives  $|T_2(\theta)|$  for a frequency higher than  $f_0$  : in this case the maximum is not at  $\theta=0^\circ$  and multiple lobes appear.

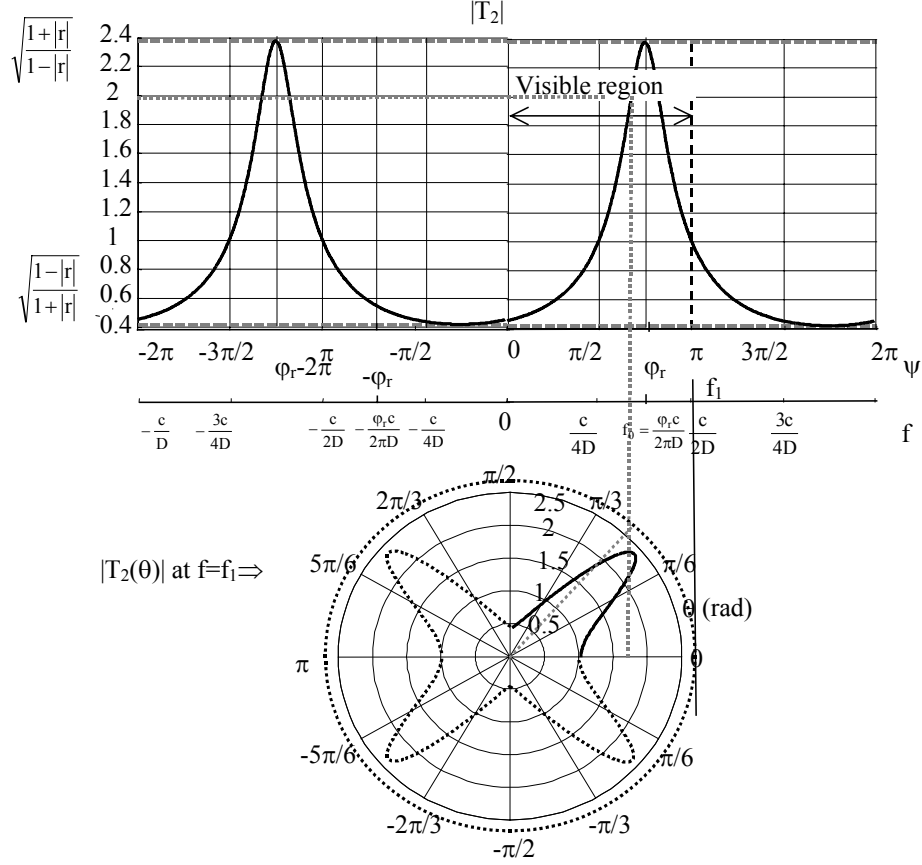


Figure 5 –  $|T_2(\psi)|$  and  $|T_2(\theta)|$  at  $f=f_1 > f_0$ .

### III. 2 Beam width versus frequency

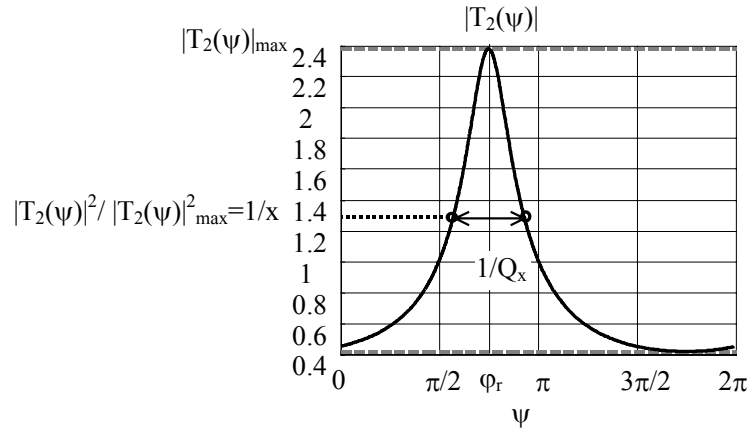


Figure 6 – Quality factor at  $1/x$ .

We call  $Q$  the quality factor of  $|T_2|$  which is obtained calculating the inverse of the bandwidth at  $-3\text{dB}$  of  $|T_2|$  multiplied by  $f_0$ .

Let us also call  $Q_x$  the quality factor of  $T_2$  obtained for a bandwidth corresponding to  $|T_2|^2 = |T_2|_{\max}^2/x$  (then  $Q_2=Q$ ) (Figure 6). In the Appendix,  $Q$  is calculated in function of  $r$  and  $Q_x$  is calculated in function of  $Q$ .

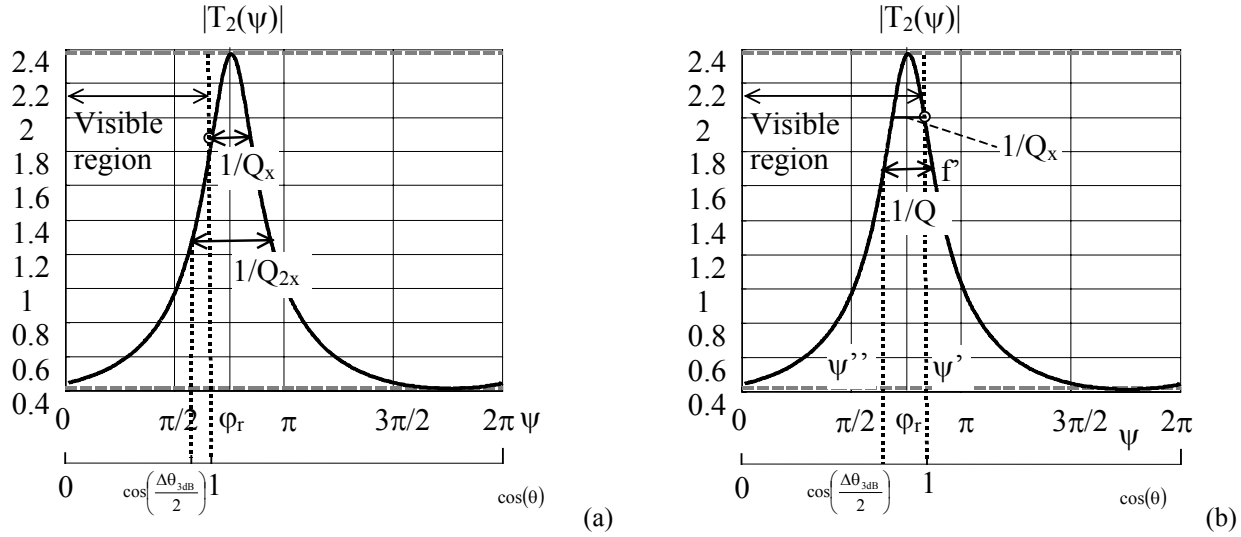


Figure 7 – Link between quality factor and beam width (a) for a frequency smaller than  $f_0$  (b) for a frequency greater than  $f_0$  and smaller than  $f'$ .

Now we will use the results presented in the Appendix to estimate the beam width  $\Delta\theta_{3\text{dB}}$  versus frequency. For a frequency smaller than  $f_0$  (see Figure 7a) we have to resolve this following equation :

$$\frac{1}{Q_{2x}} - \frac{1}{Q_x} = 2 \left( 1 - \cos \left( \frac{\Delta\theta_{3\text{dB}, f \leq f_0}}{2} \right) \right) \quad (7)$$

Considering that  $\Delta\theta_{3\text{dB}}$  is small we can write :

$$\left( \frac{\Delta\theta_{3\text{dB}, f \leq f_0}}{2} \right)^2 \approx \frac{1}{Q_{2x}} - \frac{1}{Q_x} \quad (8)$$

Then (using (22)): 
$$\Delta\theta_{3\text{dB}, f \leq f_0} \approx 2 \sqrt{\frac{\sqrt{2x-1} - \sqrt{x-1}}{Q}} \quad (9)$$

For a frequency greater than  $f_0$  but smaller than  $f'$  (see Figure 7b), this is written :

$$\frac{1}{Q_x} + \frac{1}{Q} = 2 \left( 1 - \cos \left( \frac{\Delta\theta_{3\text{dB}, f_0 \leq f \leq f'}}{2} \right) \right) \quad (10)$$

Then : 
$$\Delta\theta_{3\text{dB}, f_0 \leq f \leq f'} \approx 2 \sqrt{\frac{\sqrt{x-1} + 1}{Q}} \quad (11)$$

Note that we obtain easily the beam width at  $f_0$  in function of  $Q$ , by putting  $x=1$  in (11), we find  $\Delta\theta_{3\text{dB}, f_0} \approx \frac{2}{\sqrt{Q}}$  which is in accordance with the result in ref. [3][10]. We call this relation, the relation of performance of the structure.

This study gives a more complete result than [3][10], as all frequencies are considered. Figure 8a gives the beam width  $\Delta\theta_{3dB}$  versus the frequency ((9) and (11)) from 0 to  $f^*$  and Figure 8b gives  $\pi/\Delta\theta_{3dB}$  which we consider as an elegant evaluation of the directivity  $D_0$  at  $\theta=0^\circ$  of the structure.

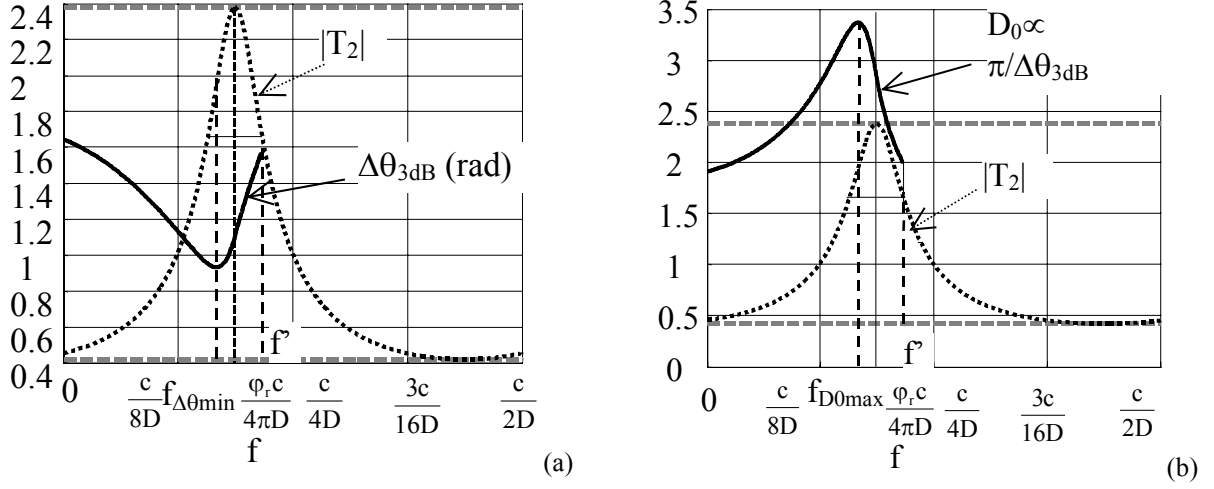


Figure 8 – (a)  $\Delta\theta_{3dB}$  and (b)  $\pi/\Delta\theta_{3dB}$  versus frequency.

Note that the minimum of  $\Delta\theta_{3dB}$  and so the maximum of directivity is not obtained at  $f_0$  but at a frequency slightly smaller than  $f_0$ . To obtain this frequency we calculate the value of  $x$  which gives the maximum of the following function :  $g(x) = \sqrt{2x-1} - \sqrt{x-1}$  (see equation (9)).

$$\text{after derivation : } g'(x) = \frac{2\sqrt{x-1} - \sqrt{2x-1}}{\sqrt{(2x-1)(x-1)}} ; \quad g'(x)=0 \Leftrightarrow x=1,5$$

$$\text{The minimum beam width is then : } \Delta\theta_{3dB, \min} \approx \sqrt{\frac{2}{Q}} \quad (12)$$

$$\text{We obtain the corresponding frequency by resolving } 2 \frac{f_0 - f_{\Delta\theta \min}}{f_0} = \frac{1}{Q \Delta\theta \min} \approx \frac{\sqrt{0.5}}{Q} = \frac{1}{\sqrt{2}Q}$$

$$\text{Finally, } f_{\Delta\theta \min} \approx f_0 \left( 1 - \frac{1}{2\sqrt{2}Q} \right) \quad (13)$$

#### IV. STRUCTURES WITH A GROUND PLANE

Let us consider now that a ground plane is placed in the same plane as the point source (in a manner that there is no radiation in the left side of the source) and there is only  $n$  PRS layers placed at a distance  $D/2$  from the source. The PRS layers are spaced from each other by a distance  $D$ . We call  $T_{n, \text{gp}}$  the transmission coefficient of the structure, whose expression becomes [6] :

$$T_{n, \text{gp}}(\psi) = \frac{t_n(\psi) \exp(-j\psi/2)}{1 - r_n(\psi) \exp(-j(\psi + \pi))} \quad (14)$$

Note that  $|t_n(\psi)|$  and  $r_n(\psi)$  are periodic with period  $\pi$  (see equations (2) and (3)). Therefore :

$$|T_{n,gp}(\psi)| = \frac{|t_n(\psi)|}{|1 - r_n(\psi)\exp(-j(\psi + \pi))|} = \frac{|t_n(\psi + \pi)|}{|1 - r_n(\psi + \pi)\exp(-j(\psi + \pi))|} = |T_{2n}(\psi + \pi)| \quad (15)$$

We deduce that the  $|T_{n,gp}(\psi)|$  curve can be obtained from the  $|T_{2n}(\psi)|$  one by  $\pi$  shifting it to the right as we can see in Figure 9. This is equivalent to add  $\pi$  to the phase of the reflection coefficient  $\varphi_r$ . Note that in our example  $\varphi_r$  is positive. We can imagine that for  $\varphi_r$  negative the first  $|T_{n,gp}(\psi)|$  pick will be higher than one.

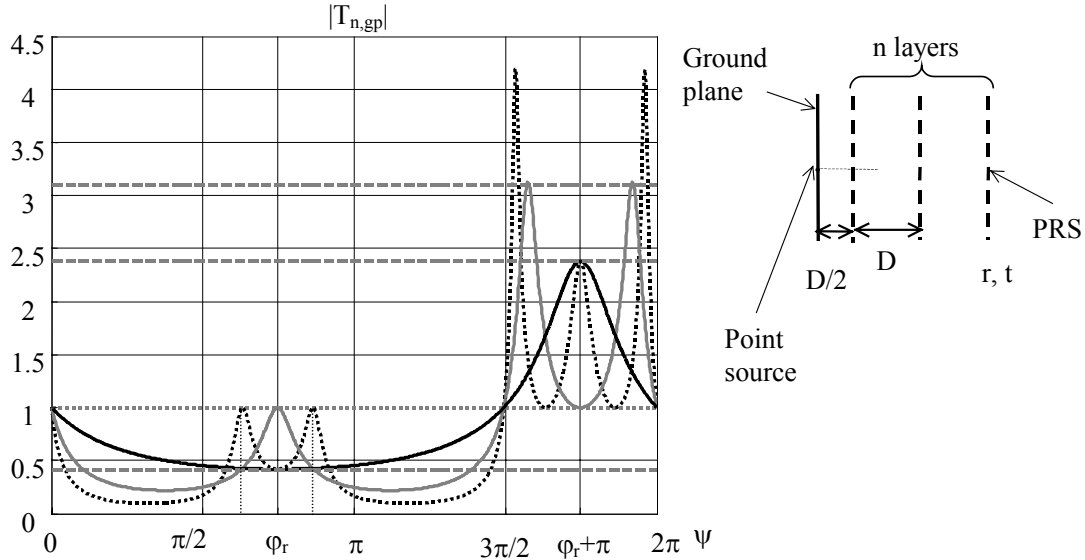


Figure 9 – Transmission coefficients  $|T_{n,pm}|$ , for  $n=1, 2$  et  $3$ , in one period  $[0, 2\pi]$ .

## V. CONCLUSION

We have presented a new and effective method to analyze 1-D Electromagnetic Band Gap based antennas. Considering Partially Reflecting Surfaces having characteristics independent of frequency and angle, we have estimate the directivity versus frequency and the frequency of maximum directivity for the simple structure composed of one layer on each side of the source. We have shown that EBG structures composed of a ground plane can be analyzed using the study of structures composed of multiple layers on each side of the source.

## REFERENCES

- [1] M. THEVENOT, C. CHEYPRE, A. REINEX and B. JECKO, “Directive Photonic Band-Gap Antennas“, IEEE Trans. On Microwave Theory and Techniques, vol. 47, no. 11, November 1999.
- [2] R. BISWAS, E. OZBAY, B. TEMELKURAN, M. BAYDINDIR, M. SIGALAS et K.-M. HO, “Exceptionally directional sources with Photonic Band-Gap crystals”, Optical Society of America, v. 18, n. 11, November 2001.
- [3] T. AKALIN, J. DANGLOT, O. VANBESIEN, and D. LIPPENS, “A highly directive Dipole Antenna Embedded in a Fabry-Perot type cavity”, IEEE Microwave and Wireless Components Letters, volume 12, No. 2, February 2002.
- [4] S. ENOCH, G. TAYEB, P. SABOUROUX, N. GUERIN and P. VINCENT, “A Metamaterial for Directive Emission”, Phys. Rev. Lett., volume 89, No. 21, November 2002.
- [5] C.A. BALANIS, “Antenna Theory : Analysis and Design”, John Wiley and Son, pp. 288, 1997.
- [6] H. BOUTAYEB, “Etude des structures périodiques planaires et conformes associées aux antennes. Application aux communications mobiles”, Ph.D dissertation, University of Rennes 1, France, 12th December 2003.
- [7] H. BOUTAYEB, K. MAHDJOUBI and A.C. TAROT, “Antenna inside PBG and Fabry-Perot cavities”, International Symposium on Antenna of Nice, JINA, November 2002.
- [8] A.P. FERESIDIS and J.C. VARDAXOGLU “High gain planar antenna using optimized partially reflective surfaces“, IEE Proc.-Microw. Antennas Propag., December 2001, vol 148, no6, pp. 345-350.
- [9] B. TEMELKURAN, E. OZBAY, J. P. KAVANAUGH, G. TUTTLE, and K.M. HO, “Resonant cavity enhanced detectors embedded in photonic crystal”, Applied Physics letters, volume 72, No. 19, 11 May 1998.



[10] H. BOUTAYEB, K. MAHDJOUBI, A.C. TAROT, “Directivité d’une structure antenne-BIP/cavité Fabry-Pérot”, JNM Symposium (Journées Nationales des Micro-ondes), Lille, France, March 2003.

## APPENDIX

To calculate Q we resolve the following equation :

$$\left| \frac{t}{1 - r \cdot \exp(-j\psi')} \right| = \frac{1}{\sqrt{2}} \sqrt{\frac{1+|r|}{1-|r|}} \quad (16)$$

which gives us  $\psi'$  and  $\psi''$  the values of  $\psi$  corresponding to the half power (see Figure 7b) :

$$\psi' = \left( \varphi_r + \arccos \left( 1 - \frac{1}{2} \frac{(1-|r|)^2}{|r|} \right) \right), \quad \psi'' = \left( \varphi_r - \arccos \left( 1 - \frac{1}{2} \frac{(1-|r|)^2}{|r|} \right) \right) \quad (17)$$

and then

$$2 \frac{\psi' - \psi''}{\psi' + \psi''} = \frac{1}{Q} = 2 \frac{\arccos \left( 1 - \frac{1}{2} \frac{(1-|r|)^2}{|r|} \right)}{\varphi_r} \quad (18)$$

This can be approximated for high value of  $|r|$  by

$$\frac{1}{Q} \approx 2 \frac{(1-|r|)}{\varphi_r \sqrt{|r|}} \quad (19)$$

By the same way, we can evaluate  $Q_x$ , resolving the following equation :

$$\left| \frac{t}{1 - r \cdot \exp(-j\psi_x)} \right| = \frac{1}{\sqrt{x}} \sqrt{\frac{1+|r|}{1-|r|}}, \quad (20)$$

We find

$$\frac{1}{Q_x} = 2 \frac{\arccos \left( 1 - \frac{1}{2} (x-1) \frac{(1-|r|)^2}{|r|} \right)}{\varphi_r} \quad (21)$$

For high value of  $|r|$ , we can write :

$$\frac{1}{Q_x} \approx \sqrt{x-1} \frac{1}{Q}, \quad \text{with} \quad x = \frac{(|T_2(\psi)|_{\max})^2}{|T_2(\psi)|^2} \quad (22)$$